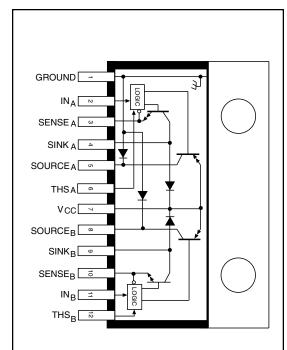
## **DUAL SOLENOID/MOTOR DRIVER** -PULSE-WIDTH MODULATED CURRENT CONTROL



Dwg. No. D-1001

#### **ABSOLUTE MAXIMUM RATINGS**

at T<sub>1</sub> ≤ +150°C

Supply Voltage, V <sub>CC</sub> 45	٧
Peak Output Current, I <sub>OUT</sub> ±4	Α
Input Voltage Range,	
V <sub>IN</sub> 0.3 V to +7.0	٧
Package Power Dissipation,	
P <sub>D</sub> <b>See Gra</b> p	h
Operating Temperature Range,	
T <sub>A</sub> 20°C to +85°	С

T<sub>S</sub>..... -55°C to +150°C NOTE: Output current rating may be limited by

duty cycle, ambient temperature, and heat sinking. Under any set of conditions, do not exceed the specified peak current and a junction temperature of +150°C.

Storage Temperature Range,

Using PWM to minimize power dissipation and maximize load efficiency, the UDN2962W dual driver is recommended for impact printer solenoids and stepper motors. It is comprised of two source/ sink driver pairs rated for continuous operation to  $\pm 3$  A. It can be connected to drive two independent loads or a single load in the fullbridge configuration. Both drivers include output clamp/flyback diodes, input gain and level shifting, a voltage regulator for singlesupply operation, and pulse-width modulated output-current control circuitry. Inputs are compatible with most TTL, DTL, LSTTL, and low-voltage CMOS or PMOS logic.

The peak output current and hysteresis for each source/sink pair is set independently. Output current, threshold voltage, and hysteresis are set by the user's selection of external resistors. At the specified outputcurrent trip level, the source driver turns off. The internal clamp diode then allows current to flow without additional input from the power supply. When the lower current trip point is reached, the source driver turns back on.

The UDN2962W is in a 12-pin single in-line power-tab package. The tab is at ground potential and needs no insulation. For highcurrent or high-frequency applications, external heat sinking may be required.

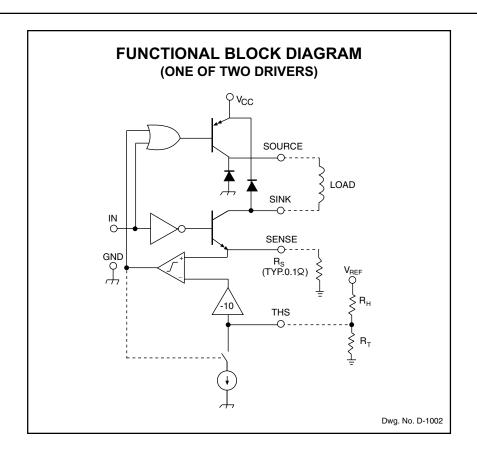
#### **FEATURES**

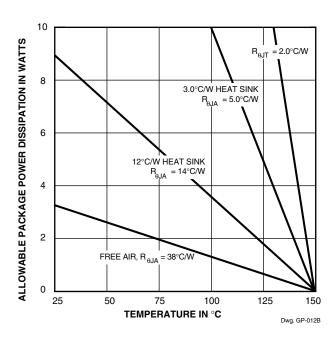
- 4 A Peak Output
- 45 V Min. Sustaining Voltage
- Internal Clamp Diodes
- TTL/PMOS/CMOS Compatible Inputs
- High-Speed Chopper

Always order by complete part number: | UDN2962W



## 2962 DUAL PWM SOLENOID/MOTOR DRIVER





### **TRUTH TABLE**

V <sub>IN</sub>	V <sub>SENSE</sub>	SOURCE DRIVER	SINK DRIVER
High	NA	Off	Off
Low	<v<sub>THS/10</v<sub>	On	On
Low	>V <sub>THS</sub> /10	Off	On



Limits

# ELECTRICAL CHARACTERISTICS at T $_{\rm A}$ = +25°C, T $_{\rm J}$ $\leq$ +150°C, V $_{\rm CC}$ = 45 V, V $_{\rm SENSE}$ = 0 V (unless otherwise noted).

Characteristic	Symbol	Test Conditions		Тур.	Max.	Units
Supply Voltage Range	V <sub>CC</sub>	Operating	20	_	45	V
Output Drivers						
Output Leakage Current	I <sub>CEX</sub>	V <sub>IN</sub> = 2.4 V, V <sub>SOURCE</sub> = 0 V	_	<-1.0	-100	μΑ
		V <sub>IN</sub> = 2.4 V, V <sub>SINK</sub> = 45 V	_	<1.0	100	μΑ
Output Saturation Voltage	V <sub>CE(SAT)</sub>	Source Drivers, I <sub>LOAD</sub> = 3.0 A		2.1	2.3	V
		Source Drivers, I <sub>LOAD</sub> = 1.0 A	_	1.7	2.0	V
		Sink Drivers, I <sub>LOAD</sub> = 3.0 A	_	1.7	2.0	V
		Sink Drivers, I <sub>LOAD</sub> = 1.0 A	_	1.1	1.3	V
Output Sustaining Voltage	V <sub>CE(sus)</sub>	I <sub>OUT</sub> = ±3.0 A, L = 3.5 mH	45	_	_	V
Output Current Regulation	$\Delta I_{OUT}$	V <sub>THS</sub> = 0.6 V to 1.0 V, L = 3.5 mH	_	_	±25	%
		V <sub>THS</sub> = 1.0 V to 2.0 V, L = 3.5 mH	_	_	±10	%
		V <sub>THS</sub> = 2.0 V to 5.0 V, L = 3.5 mH	_	_	±5.0	%
Clamp Diode Forward Voltage	V <sub>F</sub>	I <sub>F</sub> = 3.0 A	_	1.7	2.0	V
Output Rise Time	t <sub>r</sub>	I <sub>LOAD</sub> = 3.0 A, 10% to 90%, Resistive Load	_	0.5	1.0	μs
Output Fall Time	t <sub>f</sub>	I <sub>LOAD</sub> = 3.0 A, 90% to 10%, Resistive Load	_	0.5	1.0	μs
Control Logic						
Logic Input Voltage	V <sub>IN(1)</sub>		2.4	_	_	V
	V <sub>IN(0)</sub>		_	_	0.8	V
Logic Input Current	I <sub>IN(1)</sub>	V <sub>IN</sub> = 2.4 V	_	1.0	10	μА
	I <sub>IN(0)</sub>	V <sub>IN</sub> = 0.8 V	_	-20	-100	μΑ
•	I <sub>THS(ON)</sub>	$V_{THS} \ge 500 \text{ mV}, V_{SENSE} \le V_{THS}/10.5$	_	-2.0	_	μΑ
	I <sub>THS(HYS)</sub>	$V_{SENSE} \ge V_{THS}/9.5, V_{THS} = 0.6 \text{ V to } 5.0 \text{ V}$	140	200	260	μΑ
V <sub>THS</sub> /V <sub>SENSE</sub> Ratio	_	At Trip Point, V <sub>THS</sub> = 2.0 V to 5.0 V	9.5	10	10.5	_
Supply Current	I <sub>CC</sub>	V <sub>IN</sub> = 2.4 V, Outputs Off	_	8.0	12	mA
(Total Device)		V <sub>IN</sub> = 0.8 V, Outputs Open		25	40	mA
Propagation Delay Time	t <sub>pd</sub>	50% V <sub>IN</sub> to 50% V <sub>OUT</sub> , Turn Off	_	_	2.5	μs
(Resistive Load)		50% V <sub>IN</sub> to 50% V <sub>OUT</sub> , Turn On			3.0	μs
		100% V <sub>SENSE</sub> to 50% V <sub>OUT</sub> *		_	3.0	μs

<sup>\*</sup> Where  $V_{SENSE} \ge V_{THS}/9.5$ 

NOTE: Negative current is defined as coming out of (sourcing) the specified device pin.

### 2962 DUAL PWM SOLENOID/MOTOR DRIVER

## CIRCUIT DESCRIPTION AND APPLICATIONS INFORMATION

The UDN2962W high-current driver is intended for use as a free-running, pulse-width modulated solenoid driver.

**Circuit Description.** In operation, the source and sink drivers are both turned on by a low level at the input. The load current rises with time as a function of the load inductance, total circuit resistance, and supply voltage and is sensed by the external sense resistor ( $R_s$ ). When the load current reaches the trip point ( $I_{TRIP}$ ), the comparator output goes high and turns off the source driver. The actual load current will peak slightly higher than  $I_{TRIP}$  because of the internal logic and switching delays.

After the source driver is turned off, the load current continues to circulate through the sink driver and an internal ground clamp diode. The rate of current decay is a function of the load inductance and total circuit resistance.

An internal constant current sink reduces the trip point (hysteresis) until the decaying load current reaches the lower threshold, when the comparator output goes low and the source driver is again turned on. Load current is again allowed to rise to the trip point and the cycle repeats.

Maximum load current and hysteresis is determined by the user.

**Determining Maximum Load Current and Hysteresis.** Trip current ( $I_{TRIP}$ ) is determined as a function of resistance  $R_S$  and the threshold voltage,  $V_{THS}$ :

$$I_{TRIP} = \frac{V_{THS}}{10 R_S}$$

where  $V_{THS} = 10 \text{ x } V_{SENSE} = 0.6 \text{ V to } 5.0 \text{ V}.$ 

Hysteresis percentage (H) is determined by resistance  $R_{\text{H}}$  and is independent of the load current:

$$H = \frac{R_H}{50 \text{ x V}_{BEF}}$$

The chopping frequency is asynchronous and a function of the system and circuit parameters, including load inductance, supply voltage, hysteresis setting, and switching speed of the driver.

Resistance R<sub>T</sub> is determined as:

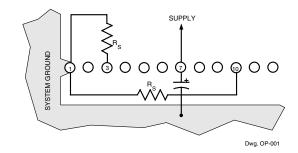
$$R_T = \frac{R_H V_{THS}}{V_{REF} - V_{THS}}$$

Note that if  $V_{THS} = V_{REF}$ , then  $R_T = \infty$ .

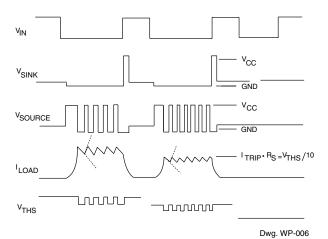
**Circuit Layout.** To prevent interaction between channels, each of the two high-level power ground returns (the low side of the sense resistors) must be returned independently to the low-level signal ground (pin 1). The circuit common (pin 1) can then be routed to the system ground.

The printed wiring board should utilize a heavy ground plane. For optimum performance, the driver should be soldered directly into the board.

The power supply  $(V_{CC})$  should be decoupled with an electrolytic capacitor  $(\ge 10 \ \mu F)$  as close as possible to pin 7.



#### **TYPICAL WAVESHAPES**

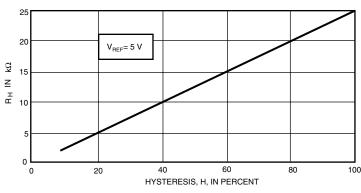




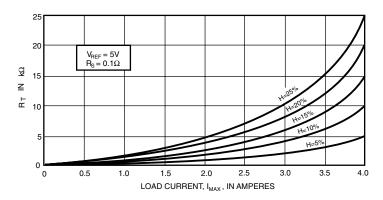
#### **APPLICATIONS INFORMATION**

## RESISTOR $R_{\rm H}$ VALUE AS A FUNCTION OF HYSTERESIS

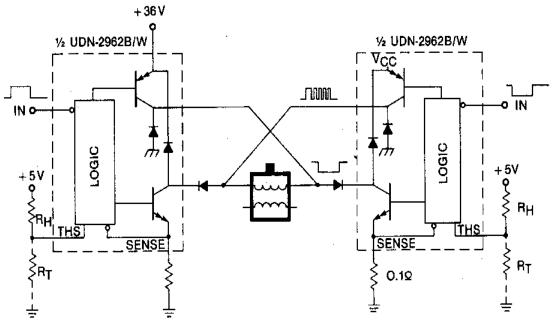
## $\begin{array}{c} {\rm RESISTOR} \; {\rm R_T} \; {\rm VALUE} \\ {\rm AS} \; {\rm A} \; {\rm FUNCTION} \; {\rm OF} \; {\rm PEAK} \; {\rm LOAD} \; {\rm CURRENT} \end{array}$



Dwg. No. A-12,417



Dwg. No. A-12,416



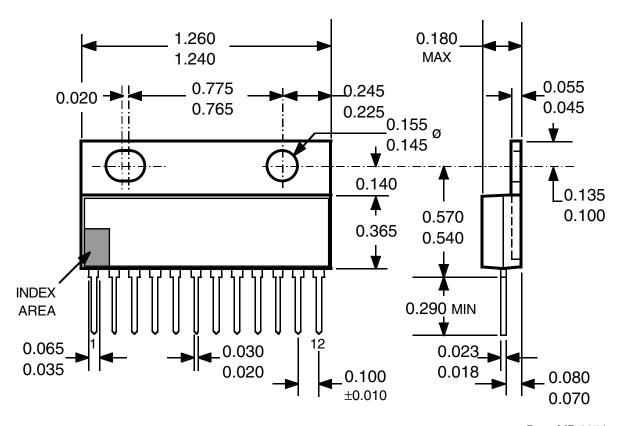
Dwg. No. D-1004

 ${\rm R}_{\rm H}$  and  ${\rm R}_{\rm T}$  determine hysteresis and peak current

NOTE: Each of the drivers includes an internal logic delay to prevent potentially destructive crossover currents within the driver during phase changes. However, never simultaneously enable both inputs in the full-bridge configurations: A destructive short-circuit to ground will result.

#### **Dimensions in Inches**

(controlling dimensions)

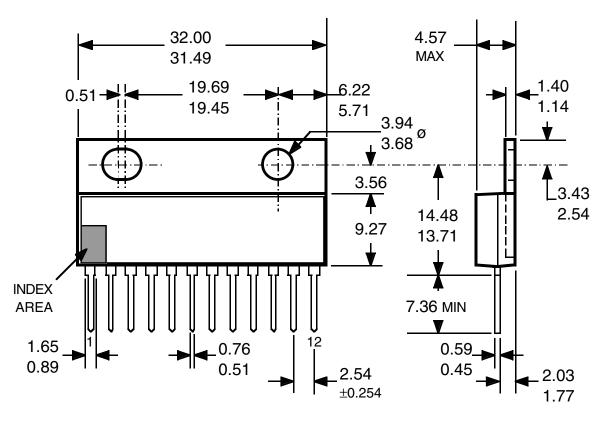


Dwg. MP-007 in

- NOTES: 1. Lead thickness is measured at seating plane or below.
  - 2. Lead spacing tolerance is non-cumulative.
  - 3. Exact body and lead configuration at vendor's option within limits shown.
  - 4. Lead gauge plane is 0.030" below seating plane.
  - 5. Supplied in standard sticks/tubes of 15 devices.

#### **Dimensions in Millimeters**

(for reference only)



Dwg. MP-007 mm

- NOTES: 1. Lead thickness is measured at seating plane or below.
  - 2. Lead spacing tolerance is non-cumulative.
  - 3. Exact body and lead configuration at vendor's option within limits shown.
  - 4. Lead gauge plane is 0.762 mm below seating plane.
  - 5. Supplied in standard sticks/tubes of 15 devices.

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### 2962 DUAL PWM SOLENOID/MOTOR DRIVER

## **MOTOR DRIVERS**

Function	Output Ratings*		Part Number <sup>†</sup>				
INTEGRATED CIRCUITS FOR BRUSHLESS DC MOTORS							
3-Phase Power MOSFET Controller	_	28 V	3933				
3-Phase Power MOSFET Controller	_	50 V	3932				
3-Phase Power MOSFET Controller	_	50 V	7600				
2-Phase Hall-Effect Sensor/Driver	400 mA	26 V	3626				
Bidirectional 3-Phase Back-EMF Controller/Driver	±600 mA	14 V	8906				
2-Phase Hall-Effect Sensor/Driver	900 mA	14 V	3625				
3-Phase Back-EMF Controller/Driver	±900 mA	14 V	8902–A				
3-Phase Controller/Drivers	±2.0 A	45 V	2936 & 2936-120				
INTEGRATED BRIDGE DRIVERS	FOR DC AND B	IPOLAR STEI	PPER MOTORS				
Dual Full Bridge with Protection & Diagnostics	±500 mA	30 V	3976				
PWM Current-Controlled Dual Full Bridge	±650 mA	30 V	3966				
PWM Current-Controlled Dual Full Bridge	±650 mA	30 V	3968				
PWM Current-Controlled Dual Full Bridge	±750 mA	45 V	2916				
PWM Current-Controlled Dual Full Bridge	±750 mA	45 V	2919				
PWM Current-Controlled Dual Full Bridge	±750 mA	45 V	6219				
PWM Current-Controlled Dual Full Bridge	±800 mA	33 V	3964				
PWM Current-Controlled Full Bridge	±1.3 A	50 V	3953				
PWM Current-Controlled Dual Full Bridge	±1.5 A	45 V	2917				
PWM Current-Controlled Microstepping Full Bridge	±1.5 A	50 V	3955				
PWM Current-Controlled Microstepping Full Bridge	±1.5 A	50 V	3957				
PWM Current-Controlled Dual DMOS Full Bridge	±1.5 A	50 V	3972				
Dual Full-Bridge Driver	±2.0 A	50 V	2998				
PWM Current-Controlled Full Bridge	±2.0 A	50 V	3952				
DMOS Full Bridge PWM Driver	±2.0 A	50 V	3958				
Dual DMOS Full Bridge	±2.5 A	50 V	3971				
UNIPOLAR STEPPER MOTOR & OTHER DRIVERS							
Voice-Coil Motor Driver	±500 mA	6 V	8932–A				
Voice-Coil Motor Driver	±800 mA	16 V	8958				
Unipolar Stepper-Motor Quad Drivers	1 A	46 V	7024 & 7029				
Unipolar Microstepper-Motor Quad Driver	1.2 A	46 V	7042				
Unipolar Stepper-Motor Translator/Driver	1.25 A	50 V	5804				
Unipolar Stepper-Motor Quad Driver	1.8 A	50 V	2540				
Unipolar Stepper-Motor Quad Driver	1.8 A	50 V	2544				
Unipolar Stepper-Motor Quad Driver	3 A	46 V	7026				
Unipolar Microstepper-Motor Quad Driver	3 A	46 V	7044				

<sup>\*</sup> Current is maximum specified test condition, voltage is maximum rating. See specification for sustaining voltage limits or over-current protection voltage limits. Negative current is defined as coming out of (sourcing) the output.

Also, see 3175, 3177, 3235, and 3275 Hall-effect sensors for use with brushless dc motors.



<sup>†</sup> Complete part number includes additional characters to indicate operating temperature range and package style.